

# LM1971 Overture™ Audio Attenuator Series Digitally Controlled 62 dB Audio Attenuator with/Mute

Check for Samples: [LM1971](#)

## FEATURES

- **3-Wire Serial Interface**
- **Mute Function**
- **Click and Pop Free Attenuation Changes**
- **8-Pin Plastic PDIP and SOIC Packages Available**

## APPLICATIONS

- **Communication Systems**
- **Cellular Phones and Pagers**
- **Personal Computer Audio Control**
- **Electronic Music (MIDI)**
- **Sound Reinforcement Systems**
- **Audio Mixing Automation**

## KEY SPECIFICATIONS

- **Total harmonic distortion 0.0008 % (typ)**
- **Frequency response > 200 kHz (–3 dB) (typ)**
- **Attenuation range (excluding mute) 62 dB (typ)**
- **Dynamic range 115 dB (typ)**
- **Mute attenuation 102 dB (typ)**

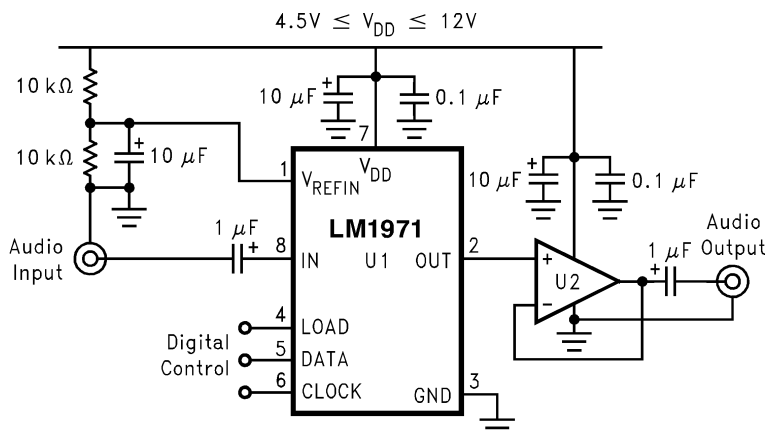
## DESCRIPTION

The LM1971 is a digitally controlled single channel audio attenuator fabricated on a CMOS process. Attenuation is variable in 1 dB steps from 0 dB to –62 dB. A mute function disconnects the input from the output, providing over 100 dB of attenuation.

The performance of the device is exhibited by its ability to change attenuation levels without audible clicks or pops. In addition, the LM1971 features a low Total Harmonic Distortion (THD) of 0.0008%, and a Dynamic Range of 115 dB, making it suitable for digital audio needs. The LM1971 is available in both 8-pin plastic PDIP or SOIC packages.

The LM1971 is controlled by a TTL/CMOS compatible 3-wire serial digital interface. The active low LOAD line enables the data input registers while the CLOCK line provides system timing. Its DATA pin receives serial data on the rising edge of each CLOCK pulse, allowing the desired attenuation setting to be selected.

## Typical Application



**Figure 1. Typical Audio Attenuator Application Circuit**



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

Overture is a trademark of dcl\_owner.

All other trademarks are the property of their respective owners.

## Connection Diagram

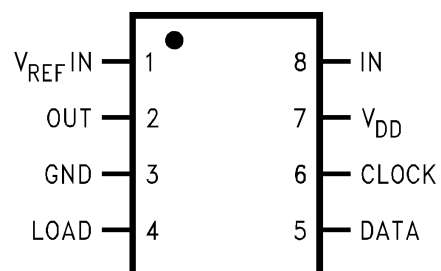


Figure 2. Dual-In-Line Plastic or Surface Mount Package- Top View



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### Absolute Maximum Ratings <sup>(1)(2)(3)</sup>

Supply Voltage, V <sub>DD</sub>			15V
Voltage at any pin			(GND -0.2V) to (V <sub>DD</sub> +0.2V)
ESD Susceptibility <sup>(4)</sup>			3000V
Soldering Information	P Package (10s)		260°C
	D Package	Vapor Phase (60s)	215°C
		Infrared (15s)	220°C
Power Dissipation <sup>(5)</sup>			150 mW
Junction Temperature			150°C
Storage Temperature			-65°C to +150°C

- (1) All voltages are measured with respect to the GND pin (pin 3), unless otherwise specified.
- (2) *Absolute Maximum Ratings* indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not specify specific performance limits. *Electrical Characteristics* state DC and AC electrical specifications under particular test conditions which ensure specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not ensured for parameters where no limit is given, however, the typical value is a good indication of device performance.
- (3) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.
- (4) Human body model, 100 pF discharged through a 1.5 k $\Omega$  resistor.
- (5) The maximum power dissipation must be derated at elevated temperatures and is dictated by  $T_{JMAX}$ ,  $\theta_{JA}$ , and the ambient temperature  $T_A$ . The maximum allowable power dissipation is  $P_D = (T_{JMAX} - T_A)/\theta_{JA}$  or the number given in the Absolute Maximum Ratings, whichever is lower. For the LM1971N and LM1971M,  $T_{JMAX} = +150^{\circ}\text{C}$ , and the typical junction-to-ambient thermal resistance,  $\theta_{JA}$ , when board mounted is 102° C/W and 167° C/W, respectively.

### Operating Ratings <sup>(1)(2)</sup>

Temperature Range	$T_{MIN} \leq T_A \leq T_{MAX}$	$-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$
Thermal Resistance	D0008A Package, $\theta_{JA}$	167°C/W
	P0008E Package, $\theta_{JA}$	102°C/W
Supply Voltage		4.5V to 12V

- (1) *Absolute Maximum Ratings* indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not specify specific performance limits. *Electrical Characteristics* state DC and AC electrical specifications under particular test conditions which ensure specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not ensured for parameters where no limit is given, however, the typical value is a good indication of device performance.
- (2) All voltages are measured with respect to the GND pin (pin 3), unless otherwise specified.

## Electrical Characteristics <sup>(1)(2)</sup>

The following specifications apply for  $V_{DD} = +12V$  ( $V_{REFIN} = +6V$ ),  $V_{IN} = 5.5 V_{pk}$ , and  $f = 1$  kHz, unless otherwise specified. Limits apply for  $T_A = 25^\circ C$ . Digital inputs are TTL and CMOS compatible.

Symbol	Parameter	Conditions	LM1971		Units (Limits)
			Typical <sup>(3)</sup>	Limit <sup>(4)</sup>	
$I_S$	Supply Current	Digital Inputs Tied to 6V	1.8	3	mA (max)
THD	Total Harmonic Distortion	$V_{IN} = 0.5V_{pk}$ @ 0 dB Attenuation	0.0008	0.003	% (max)
$e_{IN}$	Noise	Input is AC Grounded @ -12 dB Attenuation A-Weighted <sup>(5)</sup>	4.0		$\mu V$
DR	Dynamic Range	Referenced to Full Scale = $+6 V_{pk}$	115		dB
$A_M$	Mute Attenuation		102	96	dB (min)
	Attenuation Step Size Error	0 dB to -62 dB	0.009	0.2	dB (max)
	Absolute Attenuation	Attenuation @ 0 dB Attenuation @ -20 dB Attenuation @ -40 dB Attenuation @ -60 dB Attenuation @ -62 dB	0.1 -20.3 -40.5 -60.6 -62.6	0.5 -19.0 -38.0 -57.0 -59.0	dB (min) dB (min) dB (min) dB (min) dB (min)
$I_{LEAK}$	Analog Input Leakage Current	Input is AC Grounded	5.8	100	nA (max)
	Frequency Response	20 Hz–100 kHz	$\pm 0.1$		dB
$R_{IN}$	AC Input Impedance	Pin 8, $V_{IN} = 1.0 V_{pk}$ , $f = 1$ kHz	40	20 60	k $\Omega$ (min) k $\Omega$ (max)
$I_{IN}$	Input Current	@ Pins 4, 5, 6 @ $0V < V_{IN} < 5V$	1.0	100	nA (max)
$f_{CLK}$	Clock Frequency		3	2	MHz (max)
$V_{IH}$	High-Level Input Voltage	@ Pins 4, 5, 6		2.0	V (min)
$V_{IL}$	Low-Level Input Voltage	@ Pins 4, 5, 6		0.8	V (max)

- (1) *Absolute Maximum Ratings* indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not specify specific performance limits. *Electrical Characteristics* state DC and AC electrical specifications under particular test conditions which ensure specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not ensured for parameters where no limit is given, however, the typical value is a good indication of device performance.
- (2) All voltages are measured with respect to the GND pin (pin 3), unless otherwise specified.
- (3) Typicals are measured at  $25^\circ C$  and represent the parametric norm.
- (4) Limits are specifications that all parts are tested in production to meet the stated values.
- (5) Due to production test limitations, there is no limit for the Noise test. Please refer to [Figure 5](#) and [Figure 8](#) in [Typical Performance Characteristics](#).

## Pin Descriptions

**V<sub>REF</sub>IN (1):** The V<sub>REF</sub>IN pin provides the reference for the analog input signal. This pin should be biased at half of the supply voltage, V<sub>DD</sub>, as shown in [Figure 1](#) and [Figure 19](#).

**OUT (2):** The attenuated analog output signal comes from this pin.

**GND (3):** The GND pin references the digital input signals and is the lower voltage reference for the IC. Typically this pin would be labeled "V<sub>SS</sub>" but the ground reference for the digital logic input control is tied to this same point. With a higher pin-count there would generally be separate pins for these functions; V<sub>SS</sub> and Logic Ground. It is intended that the LM1971 always be operated using a single voltage supply configuration, for which pin 3 (GND) should always be at system ground. If a bipolar or split-supply configuration are desired, level shifting circuitry is needed for the digital logic control pins as they would be referenced through pin 3 which would be at the negative supply. It is highly recommended, however, that the LM1971 be used in a unipolar or single-supply configuration.

**LOAD (4):** The LOAD input accepts a TTL or CMOS level signal. This is the enable pin of the device, allowing data to be clocked in while this input is low (0V). The GND pin is the reference for this signal.

**DATA (5):** The DATA input accepts a TTL or CMOS level signal. This pin is used to accept serial data from a microcontroller that will be latched and decoded to change the channel's attenuation level. The GND pin is the reference for this signal.

**CLOCK (6):** The CLOCK input accepts a TTL or CMOS level signal. The clock input is used to load data into the internal shift register on the rising edge of the input clock waveform. The GND pin is the reference for this signal.

**V<sub>DD</sub> (7):** The positive voltage supply should be placed to this pin.

**IN (8):** The analog input signal should be placed to this pin.

## Typical Performance Characteristics

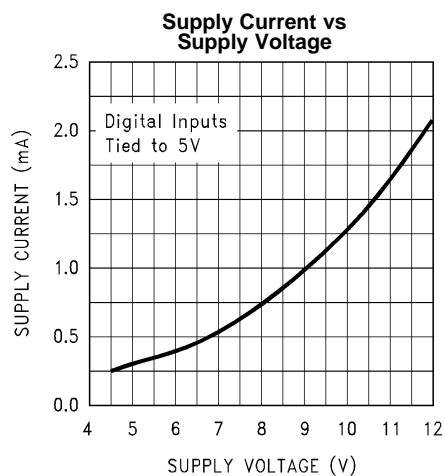


Figure 3.

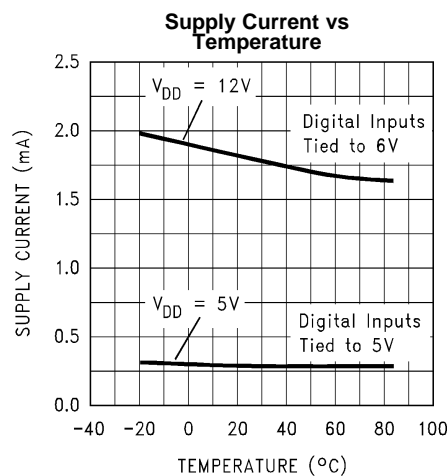


Figure 4.

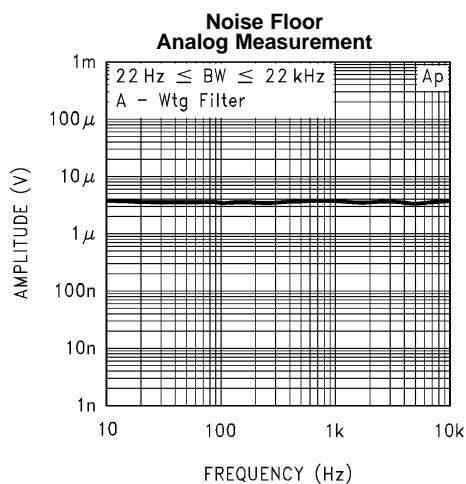


Figure 5.

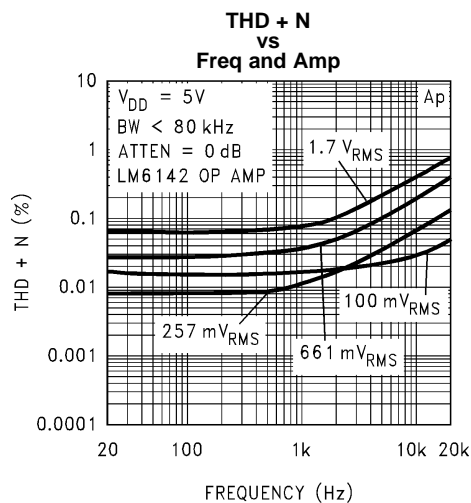


Figure 6.

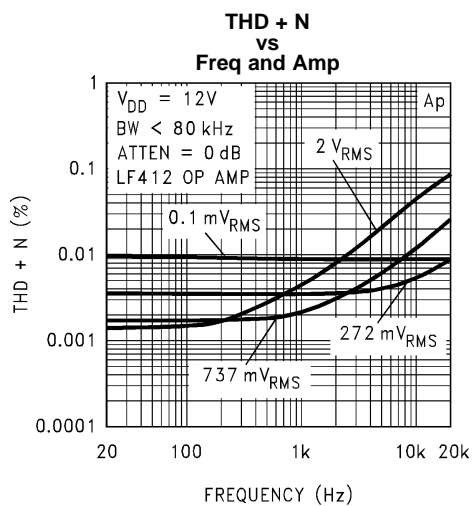


Figure 7.

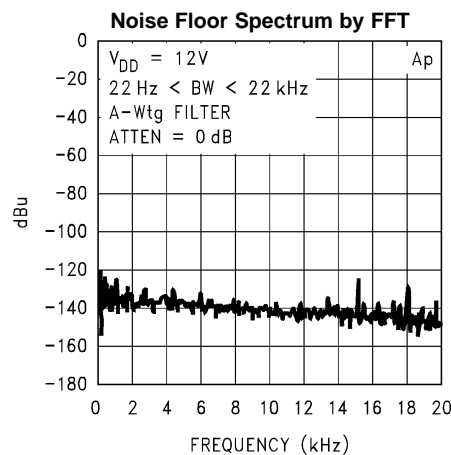


Figure 8.

## Typical Performance Characteristics (continued)

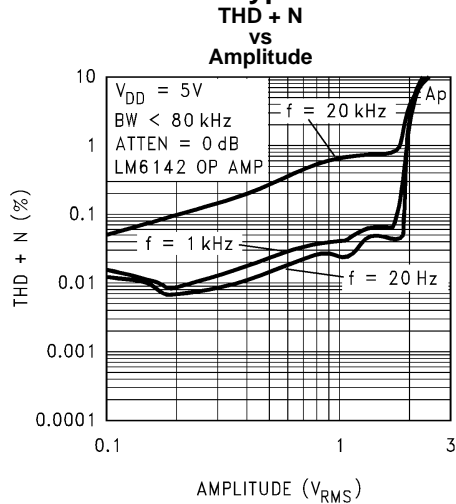


Figure 9.

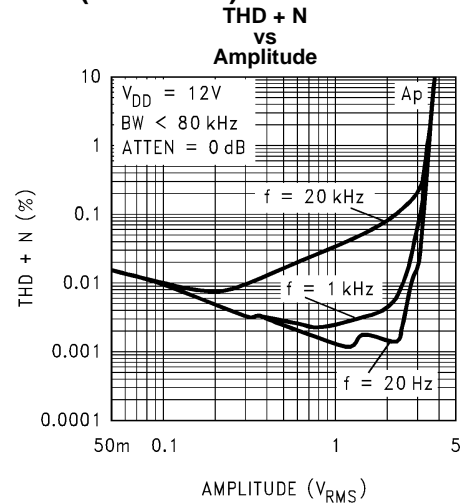


Figure 10.

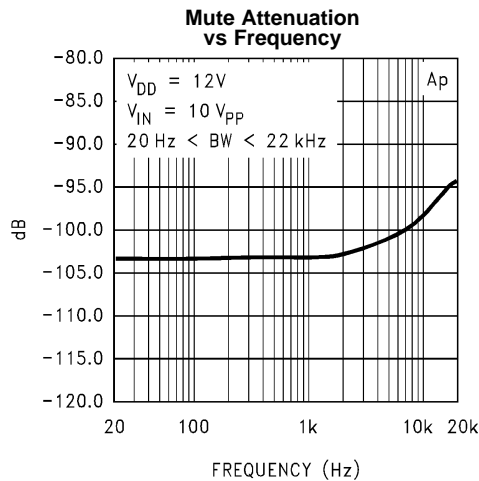


Figure 11.

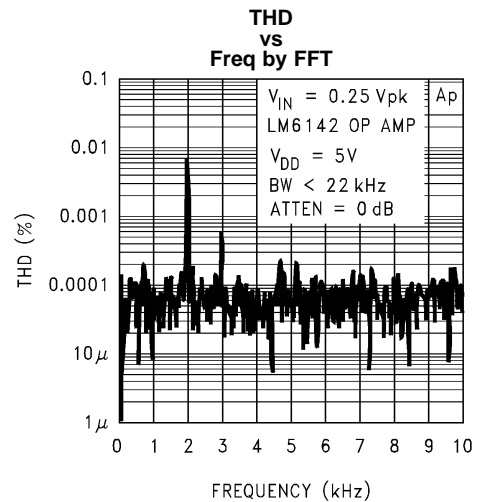


Figure 12.

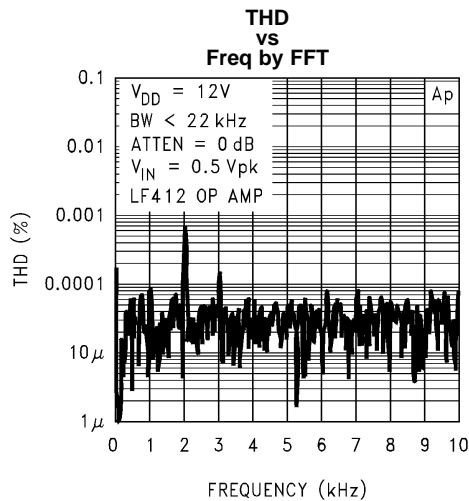


Figure 13.

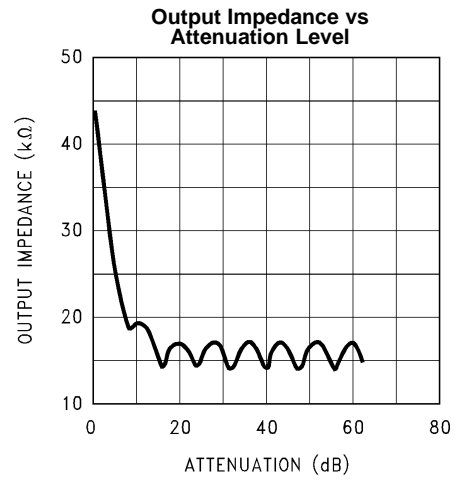


Figure 14.

## APPLICATION INFORMATION

### SERIAL DATA FORMAT

The LM1971 uses a 3-wire serial communication format that is easily controlled by a microcontroller. The timing for the 3-wire set, comprised of DATA, CLOCK, and LOAD is shown in [Figure 15](#). As depicted in [Figure 15](#), the LOAD line is to go low at least 150 ns before the rising edge of the first clock pulse and is to remain low throughout the transmission of the 16 data bits. The serial data is composed of an 8-bit address, which must always be set to 0000 0000 to select the single audio channel, and 8 bits for attenuation setting. For both address data and attenuation setting data, the MSB is sent first with the address data preceding the attenuation data. Please refer to [Figure 16](#) to confirm the serial data format transfer process.

[Table 1](#) shows the various Address and Data byte values for different attenuation settings. Note that Address bytes other than 0000 0000 are ignored.

### μPOT SYSTEM ARCHITECTURE

The μPot's digital interface is essentially a shift register where serial data is shifted in, latched, and then decoded. Once new data is shifted in, the LOAD line goes high, latching in the new data. The data is then decoded and the appropriate switch is activated to set the desired attenuation level. This process is continued each and every time an attenuation change is made. When the μPot is powered up, it is placed into the Mute mode.

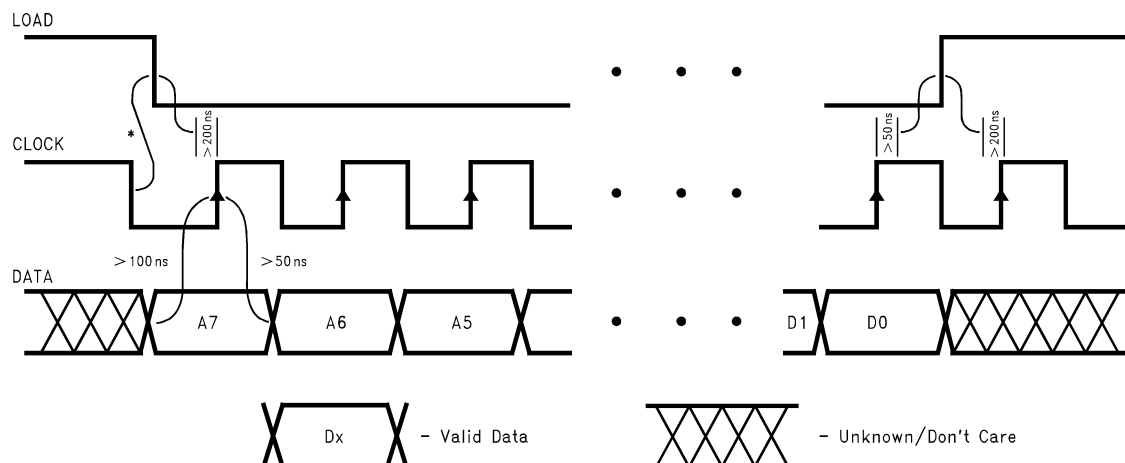
### μPOT DIGITAL COMPATIBILITY

The μPot's digital interface section is compatible with TTL or CMOS logic. The shift register inputs act upon a threshold of two diode drops above the ground level (Pin 3) or approximately 1.4V.

**Table 1. Attenuator Register Set Description**

Address Register (Byte 0)	
MSB LSB A7–A0	
0000 0000	Channel 1
0000 0001	Ignored
0000 0010	Ignored
Data Register (Byte 1)	
Contents	Attenuation (dB)
MSB LSB D7–D0	
0000 0000	0.0
0000 0001	1.0
0000 0010	2.0
0000 0011	3.0
.....	::
0001 0000	16.0
0001 0001	17.0
0001 0010	18.0
0001 0011	19.0
.....	::
0011 1101	61.0
0011 1110	62.0
0011 1111	96 (Mute)
0100 0000	96 (Mute)
.....	::
1111 1110	96 (Mute)
1111 1111	96 (Mute)





**\*Note:** Load and clock falling edges can be coincident, however, the clock falling edge cannot be delayed more than 20 ns from the falling edge of load. It is preferable that the falling edge of clock occurs before the falling edge of load.

Figure 15. Timing Diagram

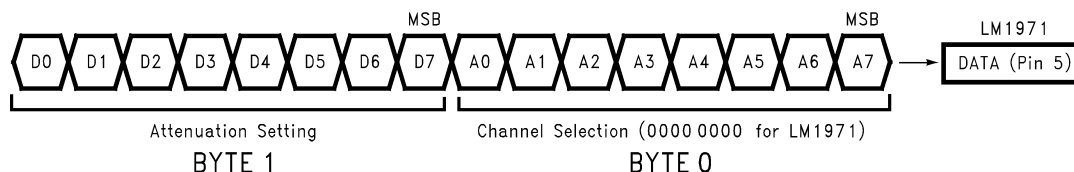


Figure 16. Serial Data Format Transfer Process

## μPOT LADDER ARCHITECTURE

The μPot contains a chain of R1/R2 resistor dividers in a ladder form, as shown in Figure 17. Each R1 is actually a series of 8 resistors, with a CMOS switch that taps into the resistor chain according to the attenuation level chosen. For any given attenuation setting, there is only one CMOS switch closed (no paralleling of ladders). The input impedance therefore remains constant, while the output impedance changes as the attenuation level changes. It is important to note that the architecture is a series of resistor dividers, and not a straight, tapped resistor, so the μPot is not a variable resistor; it is a variable voltage divider.

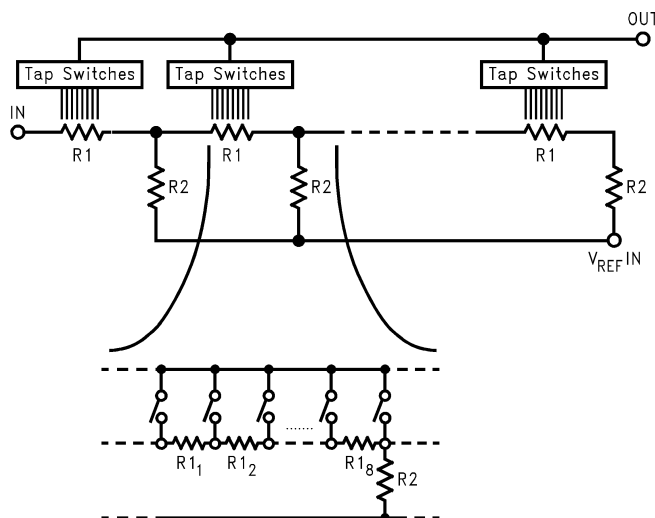
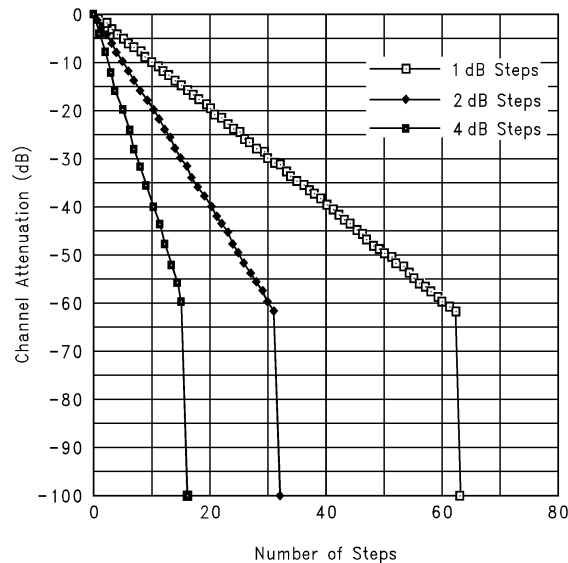


Figure 17. Resistor Ladder Architecture

## ATTENUATION STEP SCHEME

The fundamental attenuation step scheme for the LM1971 is shown in [Figure 18](#). It is also possible to obtain any integer value attenuation step through programming, in addition to the 2 dB and 4 dB steps shown in [Figure 18](#). All higher attenuation step schemes can have clickless and popless performance. Although it is possible to “skip” attenuation points by not sending all of the data, clickless and popless performance will suffer. It is highly recommended that all of the data points should be sent for each attenuation level. This ensures flawless operation and performance when making steps larger than 1 dB.



**Figure 18. LM 1971 Channel Attenuation vs Digital Step Value (1 dB, 2 dB, and 4 dB Steps)**

## INPUT IMPEDANCE

The input impedance of a  $\mu$ Pot is constant at a nominal 40 k $\Omega$ . Since the LM1971 is a single-supply operating device, it is necessary to have both input and output coupling caps as shown in [Figure 1](#). To ensure full low-frequency response, a 1  $\mu$ F coupling cap should be used.

## OUTPUT IMPEDANCE

The output impedance of a  $\mu$ Pot varies typically between 25 k $\Omega$  and 35 k $\Omega$  and changes nonlinearly with step changes. Since a  $\mu$ Pot is made up of a resistor ladder network with logarithmic attenuation, the output impedance is nonlinear. Due to this configuration, a  $\mu$ Pot cannot be considered as a linear potentiometer; it is a logarithmic attenuator.

The linearity of a  $\mu$ Pot cannot be measured directly without a buffer because the input impedance of most measurement systems is not high enough to provide the required accuracy. The lower impedance of the measurement system would load down the output and an incorrect reading would result. To prevent loading, a JFET input op amp should be used as the buffer/amplifier.

## OUTPUT BUFFERING

There are two performance issues to be aware of that are related to a  $\mu$ Pot's output stage. The first concern is to prevent audible clicks with attenuation changes, while the second is to prevent loading and subsequent linearity errors. The output stage of a  $\mu$ Pot needs to be buffered with a low input bias current op amp to keep DC shifts inaudible. Additionally, the output of  $\mu$ Pot needs to see a high impedance to keep linearity errors low.

Attenuation level changes cause changes in the output impedance of a  $\mu$ Pot. Output impedance changes in the presence of a large input bias current for a buffer/amplifier will cause a DC shift to occur. Neglecting amplifier gains and speaker sensitivities, the audibility of a DC shift is dependent upon the output impedance change times the required input bias current. As an example, a 5 k $\Omega$  impedance change times a 1  $\mu$ A bias current results in a 5 mV DC shift; a level that is barely audible without any music material in the system. An op amp with a bias current of 200 pA for the same 5 k $\Omega$  change results in an inaudible 1  $\mu$ V DC shift. Since the worst case output impedance changes are on the order of several k $\Omega$ , a bias current much less than 1  $\mu$ A is required for highest performance. In order to further quantify DC shifts, please refer to [Figure 14](#) in [Typical Performance Characteristics](#) and relate worst case impedance changes to the selected buffer/amplifier input bias current.

Without the use of a high input impedance ( $> 1 \text{ M}\Omega$ ) op amp for the buffer/amplifier, loading will occur that causes linearity errors in the signal. To ensure the highest level of performance, a JFET or CMOS input high input impedance op amp is required.

One common application that requires gain at the output of a  $\mu$ Pot is input signal volume control. Depending upon the input source material, the LM1971 provides a means of controlling the input signal level. With a supply voltage range of 4.5V to 12V, the LM1971 has the ability of controlling fairly inconsistent input source signal levels. Using an op amp with gain at the  $\mu$ Pot's output, as shown in [Figure 20](#), will also allow the system dynamic range to be increased. JFET op amps like the LF351 and the LF411 are well suited for this application. If active half-supply buffering is also desired, dual op amps like the LF353 and the LF412 could be used.

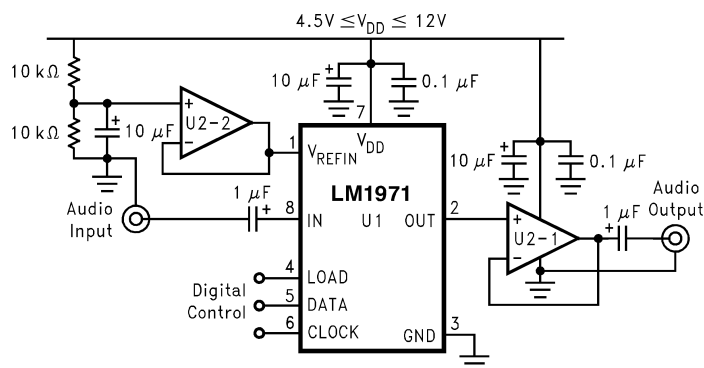
For low voltage supply applications, op amps like the CMOS LMC6041 are preferred. This part has a supply operating range from 4.5V–15.5V and also comes in a surface mount package.

## μPOT HALF-SUPPLY REFERENCING

The LM1971 operates off of a single supply, with half-supply biasing supplied at the  $V_{REF}IN$  terminal (Pin 1). The easiest and most cost effective method of providing this half-supply is a simple resistor divider and bypass capacitor network shown in [Figure 1](#). The capacitor not only stabilizes the half-supply node by “holding” the voltage nearly constant, but also decouples high frequency signals on the supply to ground. Signal feedthrough, power supply ripple and fluctuations that are not properly filtered could cause the performance of the LM1971 to be degraded.

A more stable half-supply node can be obtained by actively buffering the resistor divider network with a voltage follower as shown in [Figure 19](#). Supply fluctuations are then isolated by the high input impedance/low output impedance mismatch associated with effective filtering. Since the LM1971 is a single channel device, using a dual JFET input op amp is optimum for both output buffering and half-supply biasing.

A 10  $\mu\text{F}$  capacitor or larger is recommended for better half-supply stabilization. For added rejection of higher frequency power supply fluctuations, a smaller capacitor (0.01  $\mu\text{F}$ –0.1  $\mu\text{F}$ ) could be added in parallel to the 10  $\mu\text{F}$  capacitor.



### Figure 19. Higher Performance Active Half-Supply Buffering

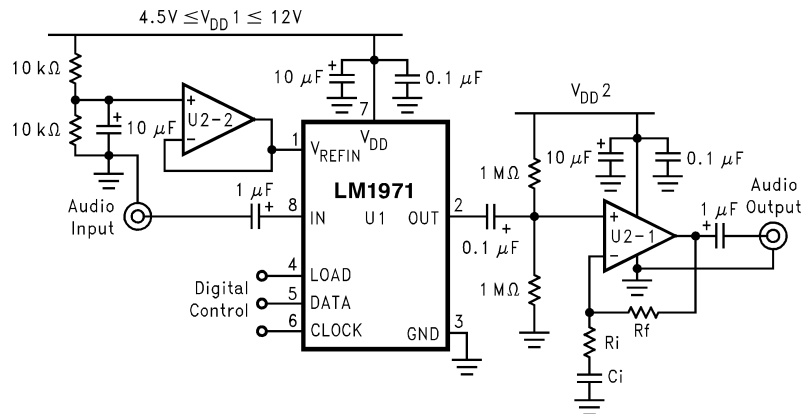


Figure 20. Active Reference with Active Gain Buffering

## LOGARITHMIC GAIN AMPLIFIER

The  $\mu$ Pot is capable of being used in the feedback loop of an op amp to create a gain controlled amplifier as shown in Figure 21. In this configuration the attenuation levels from Table 1 become gain levels with the largest possible gain value being 62 dB. For most applications, 62 dB of gain will cause signal clipping to occur. However, this can be controlled through programming. It is important to note that when in mute mode the input is disconnected from the output, thus placing the amplifier in open-loop gain state. In this mode, the amplifier will behave as a comparator. Care should be taken with the programming and design of this type of circuit. To provide the best overall performance, a high input impedance, low input bias current op amp should be used.

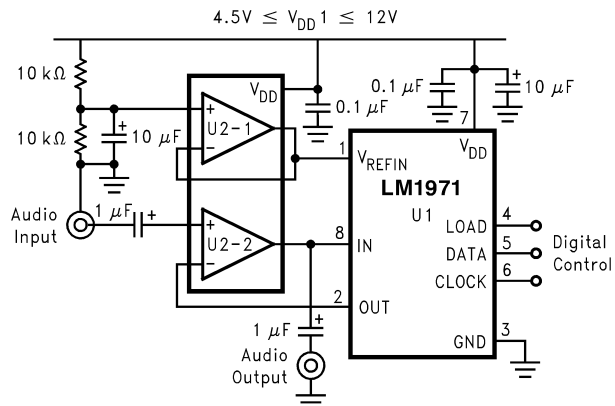


Figure 21. Logarithmic Gain Amplifier Circuit

## MUTE FUNCTION

A major feature of the LM1971 is its ability to mute the input signal to an attenuation level of 102 dB. This is accomplished internally by physically disconnecting the output from the input while also grounding the output pin through approximately 2 k $\Omega$ .

The mute function is obtained during power-up of the device or by sending any binary data of 0011 1111 and above serially to the device. The device may be placed into mute at any time during operation, allowing the designer to make the mute command accessible to the end-user.

## DC INPUTS

Although the  $\mu$ Pot was designed to be used as an attenuator for signals within the audio spectrum, it is also capable of tracking and attenuating an input DC voltage. The device will track voltages to either supply rail.

One point to remember about DC tracking is that with a buffer at the output of the  $\mu$ Pot, the resolution of DC tracking will depend upon the gain configuration of that output buffer and its supply voltage. Also, the output buffer's supply voltage does not have to be the same as the  $\mu$ Pot's supply voltage. Giving the buffer some gain can provide more resolution when tracking small DC voltages.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Top-Side Markings (4)	Samples
LM1971M	ACTIVE	SOIC	D	8	95	TBD	CU SNPB	Level-1-235C-UNLIM	-40 to 85	LM19 71M	<a href="#">Samples</a>
LM1971M/NOPB	ACTIVE	SOIC	D	8	95	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	LM19 71M	<a href="#">Samples</a>
LM1971MX	ACTIVE	SOIC	D	8	2500	TBD	CU SNPB	Level-1-235C-UNLIM	-40 to 85	LM19 71M	<a href="#">Samples</a>
LM1971MX/NOPB	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	LM19 71M	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Only one of markings shown within the brackets will appear on the physical device.

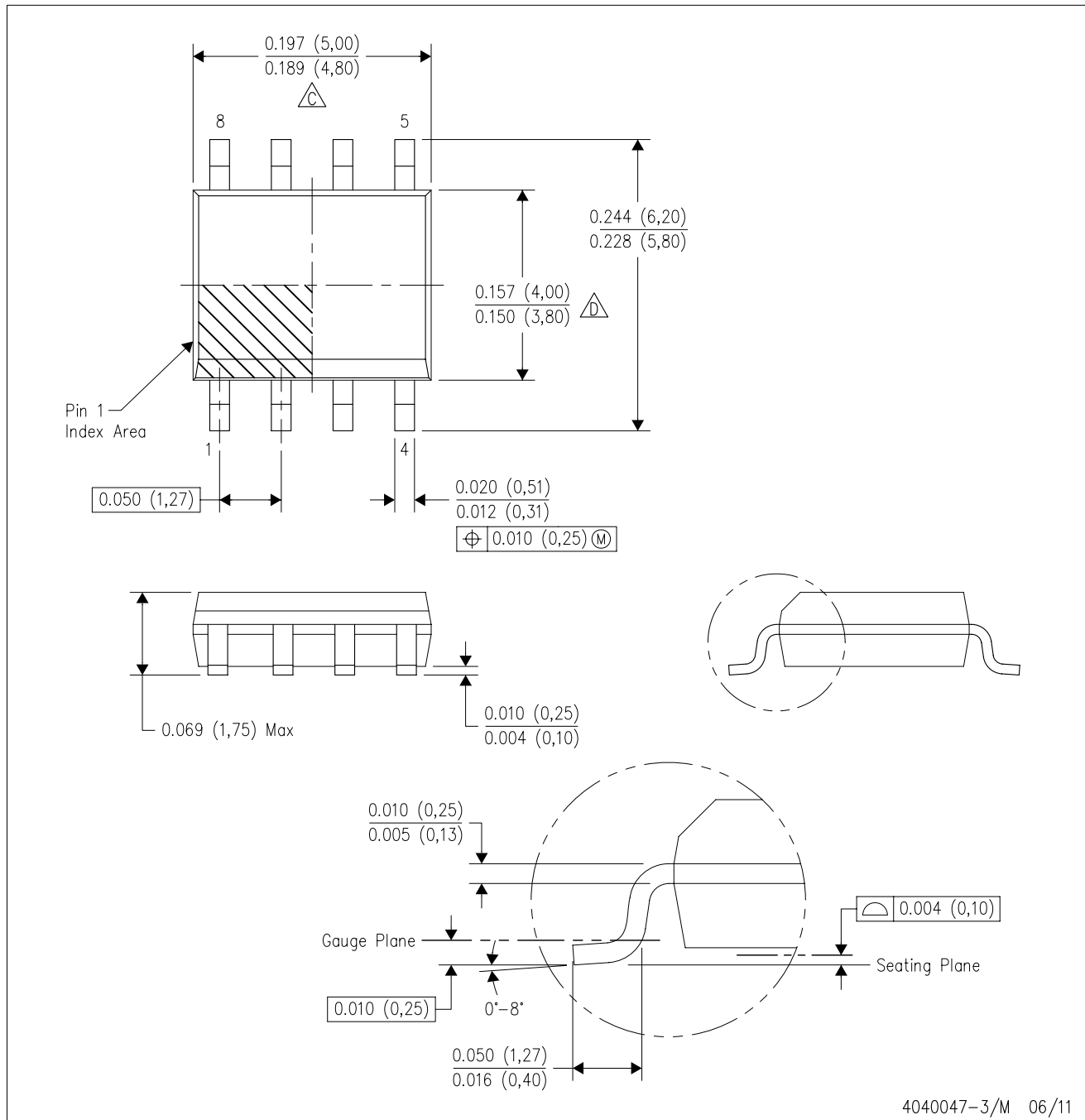
**Important Information and Disclaimer:** The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.



D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
  - B. This drawing is subject to change without notice.
  - $\triangle C$  Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
  - $\triangle D$  Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
  - E. Reference JEDEC MS-012 variation AA.



## IMPORTANT NOTICE

Texas Instruments Incorporated and its subsidiaries (TI) reserve the right to make corrections, enhancements, improvements and other changes to its semiconductor products and services per JESD46, latest issue, and to discontinue any product or service per JESD48, latest issue. Buyers should obtain the latest relevant information before placing orders and should verify that such information is current and complete. All semiconductor products (also referred to herein as "components") are sold subject to TI's terms and conditions of sale supplied at the time of order acknowledgment.

TI warrants performance of its components to the specifications applicable at the time of sale, in accordance with the warranty in TI's terms and conditions of sale of semiconductor products. Testing and other quality control techniques are used to the extent TI deems necessary to support this warranty. Except where mandated by applicable law, testing of all parameters of each component is not necessarily performed.

TI assumes no liability for applications assistance or the design of Buyers' products. Buyers are responsible for their products and applications using TI components. To minimize the risks associated with Buyers' products and applications, Buyers should provide adequate design and operating safeguards.

TI does not warrant or represent that any license, either express or implied, is granted under any patent right, copyright, mask work right, or other intellectual property right relating to any combination, machine, or process in which TI components or services are used. Information published by TI regarding third-party products or services does not constitute a license to use such products or services or a warranty or endorsement thereof. Use of such information may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

Reproduction of significant portions of TI information in TI data books or data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. TI is not responsible or liable for such altered documentation. Information of third parties may be subject to additional restrictions.

Resale of TI components or services with statements different from or beyond the parameters stated by TI for that component or service voids all express and any implied warranties for the associated TI component or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

Buyer acknowledges and agrees that it is solely responsible for compliance with all legal, regulatory and safety-related requirements concerning its products, and any use of TI components in its applications, notwithstanding any applications-related information or support that may be provided by TI. Buyer represents and agrees that it has all the necessary expertise to create and implement safeguards which anticipate dangerous consequences of failures, monitor failures and their consequences, lessen the likelihood of failures that might cause harm and take appropriate remedial actions. Buyer will fully indemnify TI and its representatives against any damages arising out of the use of any TI components in safety-critical applications.

In some cases, TI components may be promoted specifically to facilitate safety-related applications. With such components, TI's goal is to help enable customers to design and create their own end-product solutions that meet applicable functional safety standards and requirements. Nonetheless, such components are subject to these terms.

No TI components are authorized for use in FDA Class III (or similar life-critical medical equipment) unless authorized officers of the parties have executed a special agreement specifically governing such use.

Only those TI components which TI has specifically designated as military grade or "enhanced plastic" are designed and intended for use in military/aerospace applications or environments. Buyer acknowledges and agrees that any military or aerospace use of TI components which have **not** been so designated is solely at the Buyer's risk, and that Buyer is solely responsible for compliance with all legal and regulatory requirements in connection with such use.

TI has specifically designated certain components as meeting ISO/TS16949 requirements, mainly for automotive use. In any case of use of non-designated products, TI will not be responsible for any failure to meet ISO/TS16949.

### Products

Audio	<a href="http://www.ti.com/audio">www.ti.com/audio</a>
Amplifiers	<a href="http://amplifier.ti.com">amplifier.ti.com</a>
Data Converters	<a href="http://dataconverter.ti.com">dataconverter.ti.com</a>
DLP® Products	<a href="http://www.dlp.com">www.dlp.com</a>
DSP	<a href="http://dsp.ti.com">dsp.ti.com</a>
Clocks and Timers	<a href="http://www.ti.com/clocks">www.ti.com/clocks</a>
Interface	<a href="http://interface.ti.com">interface.ti.com</a>
Logic	<a href="http://logic.ti.com">logic.ti.com</a>
Power Mgmt	<a href="http://power.ti.com">power.ti.com</a>
Microcontrollers	<a href="http://microcontroller.ti.com">microcontroller.ti.com</a>
RFID	<a href="http://www.ti-rfid.com">www.ti-rfid.com</a>
OMAP Applications Processors	<a href="http://www.ti.com/omap">www.ti.com/omap</a>
Wireless Connectivity	<a href="http://www.ti.com/wirelessconnectivity">www.ti.com/wirelessconnectivity</a>

### Applications

Automotive and Transportation	<a href="http://www.ti.com/automotive">www.ti.com/automotive</a>
Communications and Telecom	<a href="http://www.ti.com/communications">www.ti.com/communications</a>
Computers and Peripherals	<a href="http://www.ti.com/computers">www.ti.com/computers</a>
Consumer Electronics	<a href="http://www.ti.com/consumer-apps">www.ti.com/consumer-apps</a>
Energy and Lighting	<a href="http://www.ti.com/energy">www.ti.com/energy</a>
Industrial	<a href="http://www.ti.com/industrial">www.ti.com/industrial</a>
Medical	<a href="http://www.ti.com/medical">www.ti.com/medical</a>
Security	<a href="http://www.ti.com/security">www.ti.com/security</a>
Space, Avionics and Defense	<a href="http://www.ti.com/space-avionics-defense">www.ti.com/space-avionics-defense</a>
Video and Imaging	<a href="http://www.ti.com/video">www.ti.com/video</a>

### TI E2E Community

[e2e.ti.com](http://e2e.ti.com)