

ON Semiconductor®

# FXMA2102 Dual Supply, 2-Bit Voltage Translator / Buffer / Repeater / Isolator for I<sup>2</sup>C Applications

#### **Features**

- Bi-Directional Interface between Any Two Levels:
   1.65 V to 5.5 V
- Direction Control not Needed
- System GPIO Resources Not Required when OE Tied to V<sub>CCA</sub>
- I<sup>2</sup>C 400 pF Buffer / Repeater
- I<sup>2</sup>C Bus Isolation
- A/B Port V<sub>OL</sub> = 175 mV (Typical), V<sub>IL</sub> = 150 mV, I<sub>OL</sub> = 6 mA
- Open-Drain Inputs / Outputs
- Accommodates Standard-Mode and Fast-Mode l<sup>2</sup>C-Bus Devices
- Supports I<sup>2</sup>C Clock Stretching & Multi-Master
- Fully Configurable: Inputs and Outputs Track V<sub>CC</sub>
- Control Input (/OE) Referenced to V<sub>CCA</sub>.
- Non-Preferential Power-Up; Either V<sub>CC</sub> May Be Powered-Up First
- Outputs Switch to 3-State if Either V<sub>CC</sub> is at GND
- Tolerant Output Enable: 5 V
- Packaged in 8-Terminal Leadless MicroPak<sup>™</sup> (1.6 mm x 1.6 mm) and Ultrathin MLP (1.2 mm x 1.4 mm)
- ESD Protection Exceeds:
  - 8 kV HBM ESD (per JESD22-A114)
  - 2 kV CDM (per JESD22-C101)

## **Description**

The FXMA2102 is a high-performance configurable dual-voltage-supply translator for bi-directional voltage translation over a wide range of input and output voltages levels.

Intended for use as a voltage translator between l<sup>2</sup>C-Bus<sup>®</sup> complaint masters and slaves.

The device is designed so that the A port tracks the  $V_{CCA}$  level and the B port tracks the  $V_{CCB}$  level. This allows for bi-directional A/B port voltage translation between any two levels from 1.65 V to 5.5 V.  $V_{CCA}$  can equal  $V_{CCB}$  from 1.65 V to 5.5 V. The OE pin is referenced to  $V_{CCA}$ .

Either  $V_{CC}$  can be pow ered-up first. Internal pow er-down control circuits place the device in 3-state if either  $V_{CC}$  is removed.

The two ports of the device have automatic direction sense capability. Either port may sense an input signal and transfer it as an output signal to the other port.

# **Ordering Information**

Part Number	Operating Temperature Range	Top Mark	Package	Packing Method
FXMA2102L8X	-40 to +85°C	XN	8-Lead MicroPak™, 1.6 mm Wide	5000 Units on
FXMA2102UMX	<del>10 10 100 0</del>	AIN	8-Lead Ultrathin MLP, 1.2 mm x 1.4 mm	Tape and Reel

# **Block Diagram**

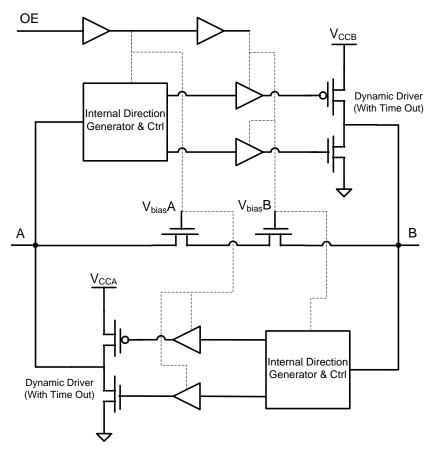
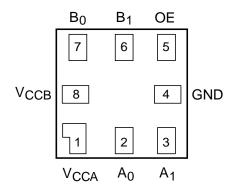
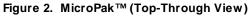


Figure 1. Block Diagram, 1 of 2 Channels

# **Pin Configuration**





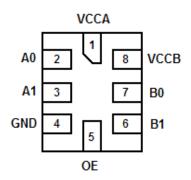


Figure 3. UMLP (Top-Through View)

## **Pin Definitions**

Pin #	Name	Description
1	V <sub>CCA</sub>	A-Side Pow er Supply
2, 3	A <sub>0</sub> , A <sub>1</sub>	A-Side Inputs or 3-State Outputs
4	GND	Ground
5	OE	Output Enable Input (Referenced to V <sub>CCA</sub> )
6, 7	B <sub>1</sub> , B <sub>0</sub>	B-Side Inputs or 3-State Outputs
8	V <sub>CCB</sub>	B-Side Power Supply

## **Truth Table**

Control	Outputs	
OE	- Outputs	
LOW Logic Level	3-State	
HIGH Logic Level	Normal Operation	

#### Note:

1. If the OE pin is driven LOW, the FXMA2102 is disabled and the  $A_0$ ,  $A_1$ ,  $B_0$ , and  $B_1$  pins (including dynamic drivers) are forced into 3-state.

## **Absolute Maximum Ratings**

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol	Parameter			Max.	Unit	
V <sub>CCA</sub> , V <sub>CCB</sub>	Supply Voltage		-0.5	7.0		
		A Port	-0.5	7.0	V	
$V_{IN}$	DC Input Voltage	B Port	-0.5	7.0	v	
		Control Input (OE)	-0.5	7.0		
		An Outputs 3-State	-0.5	7.0		
Vo	Output Voltage <sup>(2)</sup>	B <sub>n</sub> Outputs 3-State	-0.5	7.0	V	
۷O	Output Vollage	An Outputs Active	-0.5	V <sub>CCA</sub> + 0.5V	V	
		B <sub>n</sub> Outputs Active	-0.5	V <sub>CCB</sub> + 0.5V		
lıĸ	DC Input Diode Current	At V <sub>IN</sub> < 0 V		<b>-</b> 50	mA	
Юк	DC Output Diode Current	At V <sub>O</sub> < 0 V		<b>-</b> 50	mA	
IOK	DC Output Diode Current	At V <sub>O</sub> > V <sub>CC</sub>		+50	ш	
I <sub>OH</sub> / I <sub>OL</sub>	DC Output Source/Sink Curr	ent	-50	+50	mA	
lcc	DC V <sub>CC</sub> or Ground Current p	er Supply Pin		±100	mA	
P₀	Pow er Dissipation	At 400 KHz		0.129	mW	
T <sub>STG</sub>	Storage Temperature Range		-65	+150	°C	
ESD	Electrostatic Discharge	Discharge Human Body Model, JESD22-A114		8	kV	
LOD	Capability	Charged Device Mode, JESD22-C101		2	ΝV	

#### Note:

# **Recommended Operating Conditions**

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. ON Semiconductor does not recommend exceeding them or designing to Absolute Maximum Ratings.

Symbol	Parameter			Max.	Unit	
V <sub>CCA</sub> , V <sub>CCB</sub>	Pow er Supply Operating		1.65	5.50	V	
		A Port	0	5.5		
$V_{IN}$	Input Voltage	B Port	0	5.5	V	
		Control Input (OE)	0	Vcca		
0	Thermal Resistance	8-Lead MicroPak™		279.0	C°/W	
ΘJA	inema nesistance	8-Lead Ultrathin MLP		301.5	C / VV	
T <sub>A</sub>	Free Air Operating Temperature		-40	+85	°C	

#### Note:

<sup>2.</sup> lo absolute maximum rating must be observed.

<sup>3.</sup> All unused inputs and I/O pins must be held at V<sub>CCI</sub> or GND.

## **Functional Description**

## Power-Up/Power-Down Sequencing

FXM translators offer an advantage in that either  $V_{CC}$  may be powered up first. This benefit derives from the chip design. When either  $V_{CC}$  is at 0 V, outputs are in a high-impedance state. The control input (OE) is designed to track the  $V_{CCA}$  supply. A pull-down resistor tying OE to GND should be used to ensure that bus contention, excessive currents, or oscillations do not occur during power-up/power-down. The size of the pull-down resistor is based upon the current-sinking capability of the device driving the OE pin.

The recommended power-up sequence is:

- 1. Apply pow er to the first  $V_{CC}$ .
- 2. Apply pow er to the second  $V_{CC}$ .
- 3. Drive the OE input HIGH to enable the device.

The recommended power-down sequence is:

- 1. Drive OE input LOW to disable the device.
- 2. Remove power from either V<sub>CC</sub>.
- 3. Remove power from other  $V_{CC}$ .

#### Note:

Alternatively, the OE pin can be hardwired to V<sub>CCA</sub> to save GPIO pins. If OE is hardwired to V<sub>CCA</sub>, either V<sub>CC</sub> can be powered up or down first.

## **Application Circuit**

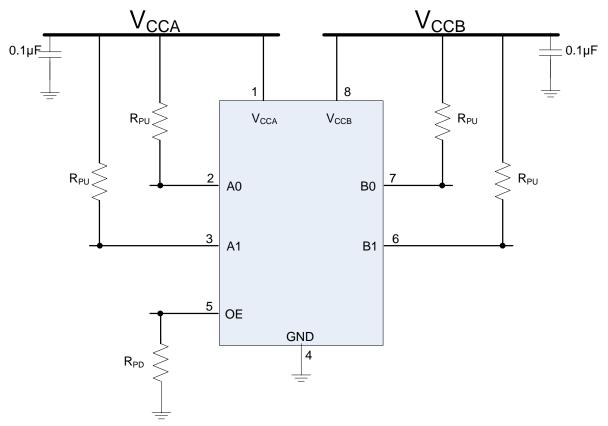


Figure 4. Application Circuit

## **Application Notes**

The FXMA2102 has open-drain I/Os and requires external pull-up resistors on the four data I/O pins, as shown in Figure 4. If a pair of data VO pins  $(A_n/B_n)$  is not used, both pins should be tied to GND (or both to V<sub>CC</sub>). In this case, pull-down or pull-up resistors are not required. The recommended values for the pull-up resistors (RPU) are 1 K $\Omega$  to 10 K $\Omega$ ; how ever, depending on the total bus capacitance, the user is free to vary the pull-up resistor value to meet the maximum I<sup>2</sup>C edge rate per the IC specification (UM10204 rev. 03, June 19, 2007). For example, the maximum edge rate (30% -70%) during fast mode (400 kbit/s) is 300 ns. If bus capacitance is approaching the maximum 400 pF, lower the RPU value to keep the rise time below 300 ns (Fast Mode). Section 7.1 of the I<sup>2</sup>C specification provides an excellent guideline for pull-up resistor sizing.

### **Theory of Operation**

The FXMA2102 is designed for high-performance level shifting and buffer / repeating in an  $l^2C$  application. Figure 1 shows that each bi-directional channel contains two series-Npassgates and two dynamic drivers. This hybrid architecture is highly beneficial in an  $l^2C$  application where auto-direction is a necessity.

For example, during the following three  $l^2C$  protocol events:

- Clock Stretching
- Slave's ACK Bit (9<sup>th</sup> bit = 0) following a Master's Write Bit (8<sup>th</sup> bit = 0)
- Clock Synchronization and Multi Master Arbitration

The bus direction needs to change from master to slave to slave to master without the occurrence of an edge. If there is an  $\mathring{\Gamma}^2C$  translator between the master and slave in these examples, the  $\mathring{\Gamma}^2C$  translator must change direction when both A and B ports are LOW. The Npassgates can accomplish this task very efficiently because, when both A and B ports are LOW, the Npassgates act as a low resistive short between the two (A and B) ports.

Due to  $^2\text{C}$ 's open-drain topology,  $^2\text{C}$  masters and slaves are not push/pull drivers. Logic LOWs are "pulled down" ( $^1\text{Isink}$ ), while logic HIGHs are "let go" (3-state). For example, when the master lets go of SCL (SCL always comes from the master), the rise time of SCL is largely determined by the RC time constant, where R =  $^1\text{RPU}$  and C = the bus capacitance. If the FXMA2102 is attached to the master [on the A port] in this example, and there is a slave on the B port, the Npassgates act as a low

resistive short between both ports until either of the port's  $V_{CC}/2$  thresholds are reached. After the RC time constant has reached the  $V_{CC}/2$  threshold of either port, the port's edge detector triggers both dynamic drivers to drive their respective ports in the LOW-to-HIGH (LH) direction, accelerating the rising edge. The resulting rise time resembles the scope shot in Figure 5. Effectively, two distinct slew rates appear in rise time. The first slew rate (slower) is the RC time constant of the bus. The second slew rate (much faster) is the dynamic driver accelerating the edge.

If both the A and B ports of the translator are HIGH, a high-impedance path exists between the A and B ports because both the Npassgates are turned off. If a master or slave device decides to pull SCL or SDA LOW, that device's driver pulls down ( $l_{\text{sink}}$ ) SCL or SDA until the edge reaches the A or B port  $V_{\text{CC}}/2$  threshold. When either the A or B port threshold is reached, the port's edge detector triggers both dynamic drivers to drive their respective ports in the HIGH-to-LOW (HL) direction, accelerating the falling edge.

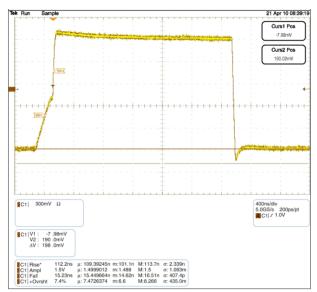


Figure 5. FXMA2102 Waveform C: 600 pF, Rpu: 2.2 K

## **Buffer/Repeater Performance**

The FXMA2102 dynamic drivers have enough current sourcing capability to drive a 400 pF capacitive bus. This is beneficial for instances when an  $\mathring{\Gamma}^2C$  buffer / repeater is required. The  $\mathring{\Gamma}^2C$  specification stipulates a maximum bus capacitance of 400 pF. If an  $\mathring{\Gamma}^2C$  segment exceeds 400 pF, an  $\mathring{\Gamma}^2C$  buffer / repeater is required to split the segment into two segments, each of which is less than 400 pF. Figure 5 is a scope shot of an FXMA2102 driving a lumped load of 600 pF. Notice the (30% - 70%) rise time is only 112 ns (RPU = 2.2 K). This is well below the maximum edge rate of 300 ns. Not only does the FXMA2102 drive 400 pF, but it also provides excellent headroom below the  $\mathring{\Gamma}^2C$  specification maximum edge rate of 300 ns.

## Vol vs. lol

The  $^{12}$ C specification mandates a maximum V<sub>IL</sub> ( $^{12}$ L of 3 mA) of V<sub>CC</sub> • 0.3 and a maximum V<sub>OL</sub> of 0.4 V. If there is a master on the A port of an  $^{12}$ C translator with a V<sub>CC</sub> of 1.65 V and a slave on the  $^{12}$ C translator B port with a V<sub>CC</sub> of 3.3 V, the maximum V<sub>IL</sub> of the master is (1.65 V x 0.3) 495 mV. The slave could legally transmit a valid logic LOW of 0.4 V to the master.

If the  ${l}^2C$  translator's channel resistance is too high, the voltage drop across the translator could present a  $V_{IL}$  to the master greater than 495 mV. To complicate matters, the  ${l}^2C$  specification states that 6 mA of  $l_{OL}$  is recommended for bus capacitances approaching 400 pF. More  $l_{OL}$  increases the voltage drop across the  ${l}^2C$  translator. The  ${l}^2C$  application benefits when  ${l}^2C$  translators exhibit low  $V_{OL}$  performance. Figure 6 depicts typical FXMA2102  $V_{OL}$  performance vs. the competition, given a 0.4  $V_{IL}$ .

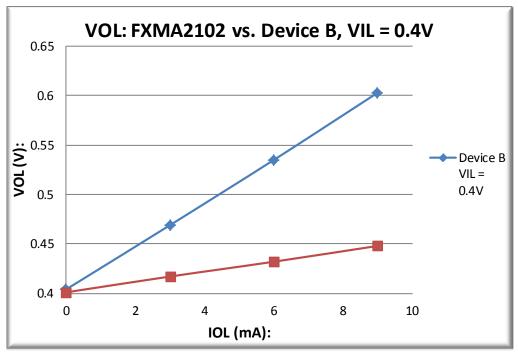


Figure 6. Vol vs. IoL

## I<sup>2</sup>C-Bus<sup>®</sup> Isolation

The FXMA2102 supports  $^{\circ}$ C-Bus isolation for the following conditions:

- Bus isolation if bus clear
- Bus isolation if either V<sub>CC</sub> goes to ground

#### **Bus Clear**

Because the I<sup>2</sup>C specification defines the minimum SCL frequency of DC, the SCL signal can be held LOW forever; how ever, this condition shuts down the I<sup>2</sup>C bus. The I<sup>2</sup>C specification refers to this condition as "Bus Clear". In Figure 7, if slave #2 holds down SCL forever, the master and slave #1 are not able to communicate, because the FXMA2102 passes the SCL stuck-LOW condition from slave #2 to slave #1 as well as the

master. However, if the OE pin is pulled LOW (disabled), both ports (A and B) are 3-stated. This results in the FXMA2102 isolating slave #2 from the master and slave #1, allowing full communication between the master and slave #1.

#### Either V<sub>CC</sub> to GND

If slave #2 is a camera that is suddenly removed from the l^C bus, resulting in  $V_{\rm CCB}$  transitioning from a valid  $V_{\rm CC}$  (1.65 V - 5.5 V) to 0 V, the FXMA2102 automatically forces SCL and SDA on both its A and B ports into 3-state. Once  $V_{\rm CCB}$  has reached 0V, full l^C communication between the master and slave #1 remains undisturbed.

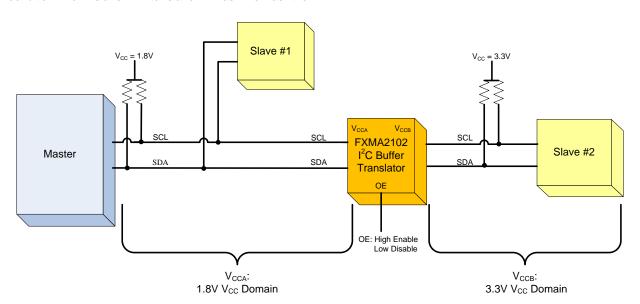


Figure 7. Bus Isolation

## **DC Electrical Characteristics**

 $T_A = -40^{\circ}C$  to +85°C.

Symbol	Parameter		Condition	V <sub>CCA</sub> (V)	V <sub>CCB</sub> (V)	Min.	Max.	Unit	
V <sub>IHA</sub>	High Level Input	h Level Input Data Inputs An		1.65-5.50	1.65-5.50	V <sub>CCA</sub> -0.4		V	
VIHA	Voltage A	Control	Input OE	1.65-5.50	1.65-5.50	0.7 x V <sub>CCA</sub>		1	
V <sub>IHB</sub>	High Level Input Voltage B	Data In	puts B <sub>n</sub>	1.65–5.50	1.65–5.50	V <sub>CCB</sub> -0.4		V	
	Low Lovel loout	Data In	puts A <sub>n</sub>	1.65-5.50	1.65-5.50		0.4	•	
$V_{ILA}$	Low Level Input Voltage A	Control	Input OE	1.65–5.50	1.65–5.50		0.3 x V <sub>CCA</sub>	V	
$V_{ILB}$	Low Level Input Voltage B	Data In	puts B <sub>n</sub>	1.65–5.50	1.65–5.50		0.4	V	
Vol	Low Level	V <sub>IL</sub> = 0	.15 V	1.65–5.50	1.65–5.50		0.4	V	
V OL	Output Voltage	loL = 6	mA	1.05-5.50	1.05-5.50		0.4	V	
IL.	Input Leakage Current		Control Input OE, V <sub>IN</sub> = V <sub>CCA</sub> or GND		1.65–5.50		±1.0	μΑ	
loff	Pow er Off	An	$V_{IN}$ or $V_O = 0$ V to 5.5 V	0	5.50		±2.0	μA	
IOFF	Leakage Current	Bn	$V_{IN}$ or $V_O = 0$ V to 5.5 V	5.50	0		±2.0	- μΑ	
		A <sub>n</sub> , B <sub>n</sub>	$V_O = 0 \text{ V to } 5.5 \text{ V},$ $OE = V_{IL}$	5.50	5.50		±2.0		
loz	3-State Output Leakage <sup>(6)</sup>	An	V <sub>O</sub> = 0 V to 5.5 V, OE = Don't Care	5.50	0		±2.0	μΑ	
		B <sub>n</sub>	V <sub>O</sub> = 0 V to 5.5 V, OE = Don't Care	0	5.50		±2.0		
lcca/в	Quiescent Supply Current <sup>(7,8)</sup>	V <sub>IN</sub> = V	$V_{IN} = V_{CCI}$ or GND, $I_O = 0$		1.65–5.50		5.0	μА	
lccz	Quiescent Supply Current <sup>(7)</sup>		$V_{IN} = V_{CCI} \text{ or GND},  I_O = 0,$ $OE = V_{IL}$		1.65–5.50		5.0	μΑ	
laar	Quiescent	t $V_{IN} = 5.5 \text{ V or GND}, I_0 = 0,$		0	1.65-5.50		-2.0	μA	
<b>I</b> CCA			Oon't Care, B <sub>n</sub> to A <sub>n</sub>	1.65-5.50	0		2.0	] ٣٨	
Іссв	Quiescent		.5 V or GND, I <sub>O</sub> = 0,	1.65–5.50	0		-2.0	μA	
-CCB	Supply Current <sup>(6)</sup>	OE = [	Oon't Care, A <sub>n</sub> to B <sub>n</sub>	0	1.65-5.50		2.0	"'	

#### Notes:

- 5. This table contains the output voltage for static conditions. Dynamic drive specifications are given in Dynamic Output Electrical Characteristics.
- 6. "Don't Care" indicates any valid logic level.
- 7. V<sub>CCI</sub> is the V<sub>CC</sub> associated with the input side.
- 8. Reflects current per supply, V<sub>CCA</sub> or V<sub>CCB</sub>.

## **Dynamic Output Electrical Characteristics**

## **Output Rise / Fall Time**

Output load:  $C_L$  = 50 pF,  $R_{PU}$  = 2.2 k $\Omega$ , push / pull driver, and  $T_A$  = -40°C to +85°C.

		V <sub>cco</sub> <sup>(10)</sup>					
Symbol	Parameter	4.5 to 5.5 V	3.0 to 3.6 V	2.3 to 2.7 V	1.65 to 1.95 V	Unit	
		Тур.	Тур.	Тур.	Тур.		
t <sub>rise</sub>	Output Rise Time; A Port, B Port <sup>(11)</sup>	3	4	5	7	ns	
t <sub>fall</sub>	Output Fall Time; A Port, B Port (12)	1	1	1	1	ns	

#### Notes:

- 9. Output rise and fall times guaranteed by design simulation and characterization; not production tested.
- 10.  $V_{CCO}$  is the  $V_{CC}$  associated with the output side.
- 11. See Figure 12.
- 12. See Figure 13.

## Maximum Data Rate<sup>(13)</sup>

Output load:  $C_L$  = 50 pF,  $R_{PU}$  = 2.2  $k\Omega$ , push / pull driver, and  $T_A$  = -40°C to +85°C.

			V	ССВ		
V <sub>CCA</sub>	Direction	4.5 to 5.5 V	3.0 to 3.6 V	2.3 to 2.7 V	1.65 to 1.95 V	Unit
		Min.	Min.	Min.	Min.	
4.5 V to 5.5 V	A to B	37	26	19	10	MHz
4.5 V 10 5.5 V	B to A	37	36	35	32	IVIITIZ
3.0 V to 3.6 V	A to B	36	25	18	10	MHz
3.0 V to 3.0 V	B to A	25	25	25	24	IVIITIZ
2.3 V to 2.7 V	A to B	35	25	18	10	MHz
2.3 V t0 2.7 V	B to A	18	18	18	17	IVIITIZ
1.65 V to 1.95 V	A to B	32	24	17	10	MHz
1.00 v to 1.90 v	B to A	10	10	10	10	1711 12

#### Note:

13. F-toggle guaranteed by design simulation; not production tested.

## **AC Characteristics**

Output Load:  $C_L = 50$  pF,  $R_{PU} = 2.2$  k $\Omega$ , and  $T_A = -40$ °C to +85°C.

		V <sub>CCB</sub>								
Symbol	l Parameter	4.5 to	5.5 V	3.0 to	3.6 V	2.3 to	2.7 V	1.65 to	1.95 V	Unit
		Тур.	Max.	Тур.	Max.	Тур.	Max.	Тур.	Max.	1
V <sub>CCA</sub> = 4.5	to 5.5 V	<u>.</u>	<b>.</b>		<b>.</b>					
tour	A to B	1	3	1	3	1	3	1	3	ns
t <sub>PLH</sub>	B to A	1	3	2	4	3	5	4	7	
4	A to B	2	4	3	5	4	6	5	7	no
t <sub>PHL</sub>	B to A	2	4	2	5	2	6	5	7	ns
to=:	OE to A	4	5	6	10	5	9	7	15	nc
t <sub>PZL</sub>	OE to B	3	5	4	7	5	8	10	15	ns
to z	OE to A	65	100	65	105	65	105	65	105	ns
t <sub>PLZ</sub>	OE to B	5	9	6	10	7	12	9	16	113
t <sub>skew</sub>	A Port, B Port <sup>(14)</sup>	0.50	1.50	0.50	1.00	0.50	1.00	0.50	1.00	ns
V <sub>CCA</sub> = 3.0	to 3.6 V	•								
4	A to B	2.0	5.0	1.5	3.0	1.5	3.0	1.5	3.0	
t <sub>PLH</sub>	B to A	1.5	3.0	1.5	4.0	2.0	6.0	3.0	9.0	ns
	A to B	2.0	4.0	2.0	4.0	2.0	5.0	3.0	5.0	ns
t <sub>PHL</sub>	B to A	2.0	4.0	2.0	4.0	2.0	5.0	3.0	5.0	
	OE to A	4.0	8.0	5.0	9.0	6.0	11.0	7.0	15.0	ns
t <sub>PZL</sub>	OE to B	4.0	8.0	6.0	9.0	8.0	11.0	10.0	14.0	
	OE to A	100	115	100	115	100	115	100	115	ns
t <sub>PLZ</sub>	OE to B	5	10	4	8	5	10	9	15	
t <sub>skew</sub>	A Port, B Port <sup>(14)</sup>	0.5	1.5	0.5	1.0	0.5	1.0	0.5	1.0	ns
V <sub>CCA</sub> = 2.3	3 to 2.7 V	•	•		•					
	A to B	2.5	5.0	2.5	5.0	2.0	4.0	1.0	3.0	
t <sub>PLH</sub>	B to A	1.5	3.0	2.0	4.0	3.0	6.0	5.0	10.0	ns
	A to B	2.0	5.0	2.0	5.0	2.0	5.0	3.0	6.0	
t <sub>PHL</sub>	B to A	2.0	5.0	2.0	5.0	2.0	5.0	3.0	6.0	ns
	OE to A	5.0	10.0	5.0	10.0	6.0	12.0	9.0	18.0	
t <sub>PZL</sub>	OE to B	4.0	8.0	4.5	9.0	5.0	10.0	9.0	18.0	ns
	OE to A	100	115	100	115	100	115	100	115	
t <sub>PLZ</sub>	OE to B	65	110	65	110	65	115	12	25	ns
t <sub>skew</sub>	A Port, B Port <sup>(14)</sup>	0.5	1.5	0.5	1.0	0.5	1.0	0.5	1.0	ns
V <sub>CCA</sub> = 1.6	55 to 1.95 V	<u> </u>								
	A to B	4	7	4	7	5	8	5	10	
t <sub>PLH</sub>	B to A	1.0	2.0	1.0	2.0	1.5	3.0	5.0	10.0	ns
	A to B	5	8	3	7	3	7	3	7	
tphL	B to A	4	8	3	7	3	7	3	7	ns
	OE to A	11	15	11	14	14	28	14	23	
t <sub>PZL</sub>	OE to B	6	14	6	12	6	12	9	16	ns
	OE to A	75	115	75	115	75	115	75	115	
t <sub>PLZ</sub>	OE to B	75	115	75	115	75	115	75	115	ns
t <sub>skew</sub>	A Port, B Port <sup>(14)</sup>	0.5	1.5	0.5	1.0	0.5	1.0	0.5	1.0	ns

#### Note:

<sup>14.</sup> Skew is the variation of propagation delay between output signals and applies only to output signals on the same port  $(A_n \text{ or } B_n)$  and switching with the same polarity (LOW-to-HIGH or HIGH-to-LOW) (see Figure 15). Skew is guaranteed, but not tested.

# Capacitance

 $T_A = +25$ °C.

Symbol	Parameter	Condition	Тур.	Unit
C <sub>IN</sub>	Input Capacitance Control Pin (OE)	$V_{CCA} = V_{CCB} = GND$	2.2	pF
C <sub>I/O</sub>	Input/Output Capacitance, A <sub>n</sub> , B <sub>n</sub>	$V_{CCA} = V_{CCB} = 5.0 \text{ V}, OE = GND, VA = VB = 5.0 \text{ V}$	13.0	pF
C <sub>pd</sub>	Pow er Dissipation Capacitance	$V_{CCA} = V_{CCB} = 5.0 \text{ V}, V_{IN} = 0 \text{ V or } V_{CC}, f = 400 \text{ KHz}$	13.5	pF

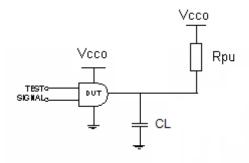


Figure 8. AC Test Circuit

Table 1. Propagation Delay Table

Test	Input Signal	Output Enable Control
t <sub>PLH</sub> , t <sub>PHL</sub>	Data Pulses	V <sub>CCA</sub>
t <sub>PZL</sub> (OE to A <sub>n</sub> , B <sub>n</sub> )	0 V	LOW to HIGH Switch
t <sub>PLZ</sub> (OE to A <sub>n</sub> , B <sub>n</sub> )	0 V	HIGH to LOW Switch

Table 2. AC Load Table

V <sub>cco</sub>	C <sub>L</sub>	$R_L$
1.8 ± 0.15 V	50 pF	2.2 kΩ
2.5 ± 0.2 V	50 pF	2.2 kΩ
3.3 ± 0.3 V	50 pF	2.2 kΩ
5.0 ± 0.5 V	50 pF	2.2 kΩ

## **Timing Diagrams**

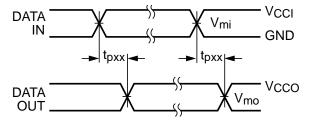


Figure 9. Waveform for Inverting and Non-Inverting Functions (15)

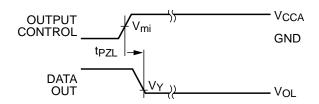


Figure 10.3-STATE Output Low Enable Time (10)

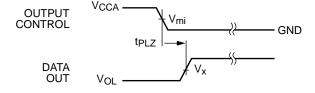
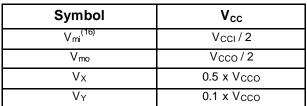


Figure 11.3-STATE Output High Enable Time<sup>(15)</sup>



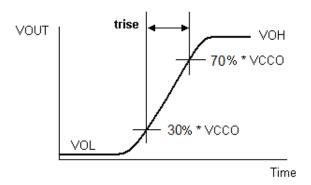


Figure 12. Active Output Rise Time

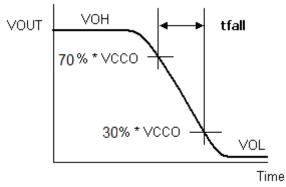


Figure 13. Active Output Fall Time

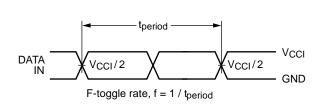
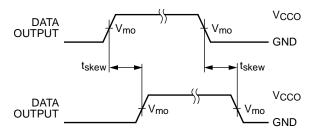


Figure 14.F-Toggle Rate



tskew = (tpHLmax - tpHLmin) or (tpLHmax - tpLHmin)

Figure 15. Output Skew Time

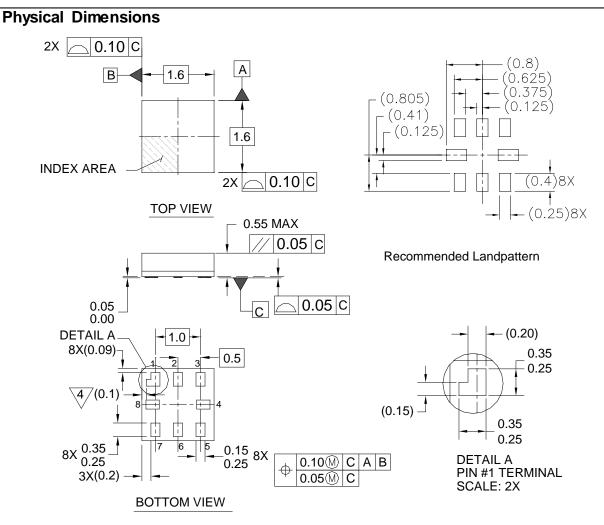
#### Notae:

- 15. Input  $t_R = t_F = 2.0$  ns, 10% to 90% at  $V_{IN} = 1.65$  V to 1.95 V; Input  $t_R = t_F = 2.0$  ns, 10% to 90% at  $V_{IN} = 2.3$ V to 2.7 V; Input  $t_R = t_F = 2.5$  ns, 10% to 90%, at  $V_{IN} = 3.0$  V to 3.6 V only; Input  $t_R = t_F = 2.5$  ns, 10% to 90%, at  $V_{IN} = 4.5$  V to 5.5V only.
- 16.  $V_{CCI} = V_{CCA}$  for control pin OE or  $V_{mi} = (V_{CCA}/2)$ .

## 8-Lead Ultrathin MLP Product-Specific Dimensions

Symbol from JEDEC MO-220	Description	NOM Value
A	Overall Height	0.55
A1	PKG Standoff	0.012
A3	Lead Thickness	0.15
b	Lead Width	0.2
D	Body Length (X)	1.4
E	Body Width (Y)	1.2
L	Lead Length	0.3
е	Lead Pitch	0.4

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#### Notes:

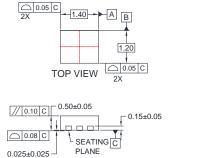
- 1. PACKAGE CONFORMS TO JEDEC MO-255 VARIATION UAAD
- 2. DIMENSIONS ARE IN MILLIMETERS
- 3. DRAWING CONFORMS TO ASME Y.14M-1994
- 4/PIN 1 FLAG, END OF PACKAGE OFFSET
- 5. DRAWING FILE NAME: MKT-MAC08AREV4

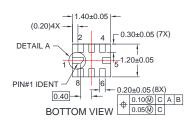
MAC08AREV4

Figure 16.8-Lead MicroPak™, 1.6 mm Wide

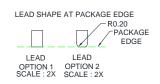
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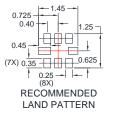
# Physical Dimensions





SIDE VIEW







#### NOTES:

- A. PACKAGE DOES NOT FULLY CONFORM TO JEDEC STANDARD.
- B. DIMENSIONS ARE IN MILLIMETERS.
- C. DIMENSIONS AND TOLERANCES PER ASME Y14.5M, 2009.
- D. LAND PATTERN RECOMMENDATION IS EXISTING INDUSTRY LAND PATTERN.
- E. DRAWING FILENAME: MKT-UMLP08Arev4.

Figure 17.8-Lead Ultrathin MLP, 1.2 mm x 1.4 mm

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